May 2006

DS15MB200 Dual 1.5 Gbps 2:1/1:2 LVDS Mux/Buffer with Pre-Emphasis



DS15MB200 Dual 1.5 Gbps 2:1/1:2 LVDS Mux/Buffer with **Pre-Emphasis General Description**

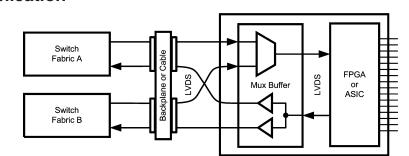
The DS15MB200 is a dual-port 2 to 1 multiplexer and 1 to 2 repeater/buffer. High-speed data paths and flow-through pinout minimize internal device jitter and simplify board layout, while pre-emphasis overcomes ISI jitter effects from lossy backplanes and cables. The differential inputs and outputs interface to LVDS or Bus LVDS signals such as those on National's 10-, 16-, and 18- bit Bus LVDS SerDes, or to CML or LVPECL signals.

The 3.3V supply, CMOS process, and robust I/O ensure high performance at low power over the entire industrial -40 to +85°C temperature range.

Features

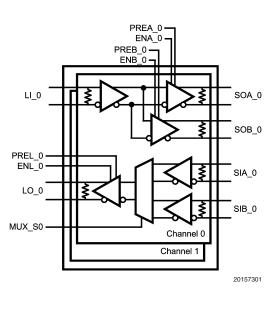
- 1.5 Gbps data rate per channel
- Configurable off/on pre-emphasis drives lossy backplanes and cables
- LVDS/BLVDS/CML/LVPECL compatible inputs, LVDS compatible outputs
- Low output skew and jitter
- On-chip 100Ω input and output termination
- 15 kV ESD protection on LVDS inputs/outputs
- Hot plug Protection
- Single 3.3V supply
- Industrial -40 to +85°C temperature range
- 48-pin LLP Package





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Block Diagram



Pin Descriptions

Pin Name	LLP Pin Number	I/O, Type	Description
SIA_0+ SIA_0–	30 29	I, LVDS	Switch A-side Channel 0 inverting and non-inverting differential inputs. LVDS, Bus LVDS, CML, or LVPECL compatible.
SIA_1+ SIA_1–	19 20	I, LVDS	Switch A-side Channel 1 inverting and non-inverting differential inputs. LVDS, Bus LVDS, CML, or LVPECL compatible.
	28 27	I, LVDS	Switch B-side Channel 0 inverting and non-inverting differential inputs. LVDS, Bus LVDS, CML, or LVPECL compatible.
SIB_1+ SIB_1-	21 22	I, LVDS	Switch B-side Channel 1 inverting and non-inverting differential inputs. LVDS, Bus LVDS, CML, or LVPECL compatible.
	DIFFEREN		
_I_0+ _I_0-	40 39	I, LVDS	Line-side Channel 0 inverting and non-inverting differential inputs. LVDS, Bus LVDS, CML, or LVPECL compatible.
LI_1+ LI_1-	9 10	I, LVDS	Line-side Channel 1 inverting and non-inverting differential inputs. LVDS, Bus LVDS, CML, or LVPECL compatible.
SWITCH S	IDE DIFFER	ENTIAL OU	ITPUTS
SOA_0+	34	O, LVDS	Switch A-side Channel 0 inverting and non-inverting differential outputs. LVDS compatible
SOA_0-	33		(Notes 1, 3).
SOA_1+	15	O, LVDS	Switch A-side Channel 1 inverting and non-inverting differential outputs. LVDS compatible
SOA_1-	16		(Notes 1, 3).
SOB_0+	32	O, LVDS	Switch B-side Channel 0 inverting and non-inverting differential outputs. LVDS compatible
SOB_0-	31		(Notes 1, 3).
SOB_1+	17	O, LVDS	Switch B-side Channel 1 inverting and non-inverting differential outputs. LVDS compatible
SOB_1-	18		(Notes 1, 3).
INE SIDE	DIFFEREN		UTS
_O_0+	42	O, LVDS	Line-side Channel 0 inverting and non-inverting differential outputs. LVDS compatible (Notes
_0_0-	41		1, 3).
_0_1+	7	O, LVDS	Line-side Channel 1 inverting and non-inverting differential outputs. LVDS compatible (Notes
_0_1-	8		1, 3).
DIGITAL C		ITERFACE	
MUX_S0	38	I, LVTTL	Mux Select Control Inputs (per channel) to select which Switch-side input, A or B, is passed
MUX_S1	11		through to the Line-side.
PREA_0	26	I, LVTTL	Output pre-emphasis control for Switch-side outputs. Each output driver on the Switch A-side
PREA_1	23		and B-side has a separate pin to control the pre-emphasis on or off.
PREB_0	25		
PREB_1	24		
PREL_0	44	I, LVTTL	Output pre-emphasis control for Line-side outputs. Each output driver on the Line A-side and
PREL_1	5		B-side has a separate pin to control the pre-emphasis on or off.
ENA_0	36	I, LVTTL	Output Enable Control for Switch A-side and B-side outputs. Each output driver on the A-side
ENA_1	13		and B-side has a separate enable pin.
ENB_0	35		
	14		
ENB_1			
ENB_1 ENL_0	45	I, LVTTL	Output Enable Control for The Line-side outputs. Each output driver on the Line-side has a

DS15MB200

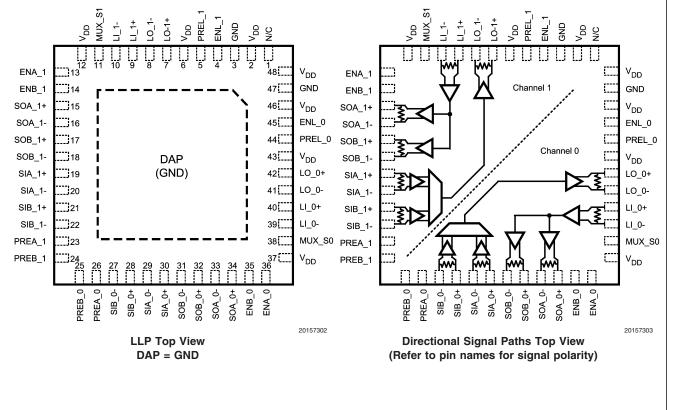
Pin	LLP Pin		Description	
Name Number		I/O, Type	Description	
POWER	•	•		
V _{DD}	2, 6, 12,	I, Power	$V_{DD} = 3.3V \pm 0.3V.$	
	37, 43,			
	46, 48			
GND	3, 47	I, Power	Ground reference for LVDS and CMOS circuitry.	
	(Note 2)		For the LLP package, the DAP is used as the primary GND connection to the device. The	
			DAP is the exposed metal contact at the bottom of the LLP-48 package. It should be	
			connected to the ground plane with at least 4 vias for optimal AC and thermal performance.	

Note 1: For interfacing LVDS outputs to CML or LVPECL compatible inputs, refer to the applications section of this datasheet (planned).

Note 2: Note that the DAP on the backside of the LLP package is the primary GND connection for the device when using the LLP package.

Note 3: The LVDS outputs do not support a multidrop (BLVDS) environment. The LVDS output characteristics of the DS15MB200 device have been optimized for point-to-point backplane and cable applications.

Connection Diagrams



Output Characteristics

The output characteristics of the DS15MB200 have been optimized for point-to-point backplane and cable applications, and are not intended for multipoint or multidrop signaling.

A 100 Ω output (source) termination resistor is incorporated in the device to eliminate the need for an external resistor, providing excellent drive characteristics by locating the source termination as close to the output as physically possible.

Pre-Emphasis Controls

The pre-emphasis is used to compensate for long or lossy transmission media. Separate pins are provided for each output to minimize power consumption. Pre-emphasis is programmable to be off or on per the Pre-emphasis Control Table.

PREx_n (Note 4)	Output Pre-Emphasis			
0	0%			
1	100%			

Note 4: Applies to PREA_0, PREA_1, PREB_0, PREB_1, PREL_0, PREL_1

Multiplexer Truth Table (Note 5)

Data	Inputs	Contro	Output	
SIA_0	SIB_0	MUX_S0	ENL_0	LO_0
Х	valid	0	1	SIB_0
valid	Х	1	1	SIA_0
Х	Х	Х	0	Z

X = Don't Care Z = High Impedance (TRI-STATE)

Repeater/Buffer Truth Table (Note 5)

•				, ,
Data Input	Contro	l Inputs	Out	puts
•		•		-
LI_0	ENA_0	ENB_0	SOA_0	SOB_0
Х	0	0	Z	Z
valid	0	1	Z	LI_0
valid	1	0	LI_0	Z
valid	1	1	LI_0	LI_0

X = Don't Care

Z = High Impedance (TRI-STATE)

Note 5: Same functionality for channel 1

Absolute Maximum Ratings (Note 6)

	-
Supply Voltage (V _{DD})	-0.3V to +4.0V
CMOS Input Voltage	-0.3V to (V _{DD} +0.3V)
LVDS Receiver Input Voltage	
(Note 7)	-0.3V to (V _{DD} +0.3V)
LVDS Driver Output Voltage	-0.3V to (V _{DD} +0.3V)
LVDS Output Short Circuit Currer	nt +40 mA
Junction Temperature	+150°C
Storage Temperature	–65°C to +150°C
Lead Temperature (Solder, 4sec)	260°C
Max Pkg Power Capacity @ 25°C	5.2W
Thermal Resistance (θ_{JA})	24°C/W
Package Derating above +25°C	41.7mW/°C
ESD Last Passing Voltage	
HBM, 1.5kΩ, 100pF	8kV
LVDS pins to GND only	15kV

EIAJ, 0Ω, 200pF CDM 250V 1000V

Recommended Operating Conditions

Supply Voltage (V _{CC})	3.0V to 3.6V
Input Voltage (V _I) (Note 7)	0V to $V_{\rm CC}$
Output Voltage (V _O)	0V to $V_{\rm CC}$
Operating Temperature (T _A)	
Industrial	–40°C to +85°C

Note 6: Absolute maximum ratings are those values beyond which damage to the device may occur. The databook specifications should be met, without exception, to ensure that the system design is reliable over its power supply, temperature, and output/input loading variables. National does not recommend operation of products outside of recommended operation conditions. **Note 7:** V_{ID} max < 2.4V

Electrical Characteristics

Over recommended operating supply and temperature ranges unless other specified.

Symbol	Parameter	Conditions	Min	Typ (Note 8)	Max	Units
LVTTL DO	SPECIFICATIONS (MUX_Sn, PRI	EA_n, PREB_n, PREL_n, ENA_n, ENB_n,	ENL_n)			
V _{IH}	High Level Input Voltage		2.0		V_{DD}	V
V _{IL}	Low Level Input Voltage		GND		0.8	V
I _{IH}	High Level Input Current	$V_{IN} = V_{DD} = V_{DDMAX}$	-10		+10	μA
I _{IHR}	High Level Input Current	PREA_n, PREB_n, PREL_n	40		200	μA
I _{IL}	Low Level Input Current	$V_{IN} = V_{SS}, V_{DD} = V_{DDMAX}$	-10		+10	μA
C _{IN1}	Input Capacitance	Any Digital Input Pin to V _{SS}		2.0		pF
C _{OUT1}	Output Capacitance	Any Digital Output Pin to V _{SS}		4.0		pF
/ _{CL}	Input Clamp Voltage	$I_{CL} = -18 \text{ mA}$	-1.5	-0.8		V
LVDS INP	UT DC SPECIFICATIONS (SIA±, S	SIB±, LI±)				
V _{TH}	Differential Input High Threshold (Note 9)	$V_{CM} = 0.8V$ or 1.2V or 3.55V, $V_{DD} = 3.6V$		0	100	mV
V _{TL}	Differential Input Low Threshold (Note 9)	$V_{CM} = 0.8V \text{ or } 1.2V \text{ or } 3.55V,$ $V_{DD} = 3.6V$	-100	0		mV
V _{ID}	Differential Input Voltage	$V_{CM} = 0.8V$ to 3.55V, $V_{DD} = 3.6V$	100		2400	mV
V _{CMR}	Common Mode Voltage Range	$V_{ID} = 150 \text{ mV}, V_{DD} = 3.6 \text{V}$	0.05		3.55	V
C _{IN2}	Input Capacitance	IN+ or IN– to V _{SS}		2.0		pF
I _{IN}	Input Current	$V_{IN} = 3.6V, V_{DD} = V_{DDMAX} \text{ or } 0V$	-15		+15	μA
		$V_{IN} = 0V, V_{DD} = V_{DDMAX} \text{ or } 0V$	-15		+15	μA
LVDS OU	TPUT DC SPECIFICATIONS (SOA	_n±, SOB_n±, LO_n±)				
V _{OD}	Differential Output Voltage, 0% Pre-emphasis (Note 9)	R_L is the internal 100 Ω between OUT+ and OUT-	250	360	500	mV
ΔV _{OD}	Change in V _{OD} between Complementary States		-35		35	mV
V _{os}	Offset Voltage (Note 10)		1.05	1.22	1.475	V
ΔV _{OS}	Change in V _{OS} between Complementary States		-35		35	mV
os	Output Short Circuit Current	OUT+ or OUT- Short to GND		-21	-40	mA
C _{OUT2}	Output Capacitance	OUT+ or OUT- to GND when TRI-STATE		4.0		pF

DS15MB200

Symbol	Parameter	Conditions	Min	Typ (Note 8)	Max	Units
SUPPLY	CURRENT (Static)	· · · · · ·		-11		1
I _{cc}	Supply Current	All inputs and outputs enabled and active, terminated with external load of 100Ω between OUT+ and OUT		225	275	mA
I _{CCZ}	Supply Current - Powerdown Mode	ENA_0 = ENB_0 = ENL_0 = ENA_1 = ENB_1 = ENL_1 = L		0.6	4.0	
SWITCHI	NG CHARACTERISTICS—LVDS O	UTPUTS				
t _{LHT}	Differential Low to High Transition Time	Use an alternating 1 and 0 pattern at 200 Mb/s, measure between 20% and		170	250	ps
t _{HLT}	Differential High to Low Transition Time	80% of V _{OD} . (Note 15)		170	250	ps
t _{PLHD}	Differential Low to High Propagation Delay	Use an alternating 1 and 0 pattern at 200 Mb/s, measure at 50% V _{OD}		1.0	2.5	ns
t _{PHLD}	Differential High to Low Propagation Delay	between input to output.		1.0	2.5	ns
t _{SKD1}	Pulse Skew	It _{PLHD} -t _{PHLD} I (Note 15)		25	75	ps
t _{skcc}	Output Channel to Channel Skew	Difference in propagation delay (t _{PLHD} or t _{PHLD}) among all output channels. (Note 15)		50	115	ps
t _{JIT}	Jitter (0% Pre-emphasis) (Note 11)	RJ - Alternating 1 and 0 at 750MHz (Note 12)		1.1	1.5	psrms
		DJ - K28.5 Pattern, 1.5 Gbps (Note 13)		20	34	psp-p
		TJ - PRBS 2 ⁷ -1 Pattern, 1.5 Gbps (Note 14)		14	28	psp-p
t _{on}	LVDS Output Enable Time	Time from ENA_n, ENB_n, or ENL_n to OUT± change from TRI-STATE to active.		0.5	1.5	μs
t _{on2}	LVDS Output Enable Time from Powerdown Mode	Time from ENA_n, ENB_n, or ENL_n to OUT± change from Powerdown Mode to active.		10	20	μs
t _{OFF}	LVDS Output Disable Time	Time from ENA_n, ENB_n, or ENL_n to OUT± change from active to TRI-STATE or Powerdown mode.			12	ns

Note 8: Typical parameters are measured at V_{DD} = 3.3V, T_A = 25°C. They are for reference purposes, and are not production-tested.

Note 9: Differential output voltage V_{OD} is defined as ABS(OUT+-OUT-). Differential input voltage V_{ID} is defined as ABS(IN+-IN-).

Note 10: Output offset voltage V_{OS} is defined as the average of the LVDS single-ended output voltages at logic high and logic low states.

Note 11: Jitter is not production tested, but guaranteed through characterization on a sample basis.

Note 12: Random Jitter, or RJ, is measured RMS with a histogram including 1500 histogram window hits. The input voltage = V_{ID} = 500mV, 50% duty cycle at 750MHz, $t_r = t_f = 50$ ps (20% to 80%).

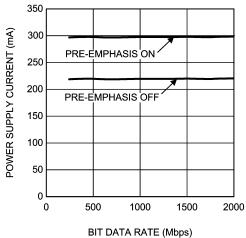
Note 13: Deterministic Jitter, or D_J, is measured to a histogram mean with a sample size of 350 hits. Stimulus and fixture jitter have been subtracted. The input voltage = V_{ID} = 500mV, K28.5 pattern at 1.5 Gbps, $t_r = t_f = 50$ ps (20% to 80%). The K28.5 pattern is repeating bit streams of (001111010 1100000101).

Note 14: Total Jitter, or T_J, is measured peak to peak with a histogram including 3500 window hits. Stimulus and fixture jitter have been subtracted. The input voltage = V_{ID} = 500mV, $2^{7.1}$ PRBS pattern at 1.5 Gbps, $t_r = t_f = 50ps$ (20% to 80%).

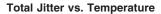
Note 15: Not production tested. Guaranteed by statistical analysis on a sample basis at the time of characterization.

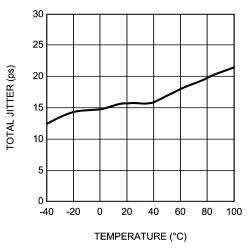
Typical Performance Characteristics

Power Supply Current vs. Bit Data Rate



20157320 Dynamic power supply current was measured with all channels active and toggling at the bit data rate. Data pattern has no effect on the power consumption. V_{DD} = 3.3V, T_A = +25 $^\circ C,~V_{ID}$ = 0.5V, V_{CM} = 1.2V

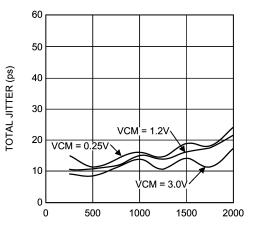




20157322 Total Jitter measured at 0V differential while running a PRBS 27-1 pattern with one channel active, all other channels are disabled. V_{DD} = 3.3V, V_{ID} = 0.5V, V_{CM} = 1.2V, 1.5 Gbps data rate, pre-emphasis off.



Total Jitter vs. Bit Data Rate



BIT DATA RATE (Mbps)

\$20157321\$ Total Jitter measured at 0V differential while running a PRBS $2^{7\text{-}1}$ pattern with one channel active, all other channels are disabled. V_{DD} = 3.3V, T_A = +25°C, V_{ID} = 0.5V, pre-emphasis off.

TRI-STATE and Powerdown Modes

The DS15MB200 has output enable control on each of the six onboard LVDS output drivers. This control allows each output individually to be placed in a low power TRI-STATE mode while the device remains active, and is useful to reduce power consumption on unused channels. In TRI-STATE mode, some outputs may remain active while some are in TRI-STATE.

When all six of the output enables (all drivers on both channels) are deasserted (LOW), then the device enters a Powerdown mode that consumes only 0.5mA (typical) of supply current. In this mode, the entire device is essentially powered off, including all receiver inputs, output drivers and internal bandgap reference generators. When returning to active mode from Powerdown mode, there is a delay until valid data is presented at the outputs because of the ramp to power up the internal bandgap reference generators.

Any single output enable that remains active will hold the device in active mode even if the other five outputs are in TRI-STATE.

When in Powerdown mode, any output enable that becomes active will wake up the device back into active mode, even if the other five outputs are in TRI-STATE.

Input Failsafe Biasing

External pull up and pull down resistors may be used to provide enough of an offset to enable an input failsafe under open-circuit conditions. This configuration ties the positive LVDS input pin to VDD thru a pull up resistor and the negative LVDS input pin is tied to GND by a pull down resistor. The pull up and pull down resistors should be in the 5k Ω to 15k Ω range to minimize loading and waveform distortion to the driver. The common-mode bias point ideally should be set to approximately 1.2V (less than 1.75V) to be compatible with the internal circuitry. Please refer to application note AN-1194, "Failsafe Biasing of LVDS Interfaces" for more information.

Interfacing LVPECL to LVDS

An LVPECL driver consists of a differential pair with coupled emitters connected to GND via a current source. This drives a pair of emitter-followers that require a 50 ohm to $V_{\rm CC}$ -2.0 load. A modern LVPECL driver will typically include the termination scheme within the device for the emitter follower. If the driver does not include the load, then an external scheme must be used. The 1.3 V supply is usually not readily available on a PCB, therefore, a load scheme without a unique power supply requirement may be used.

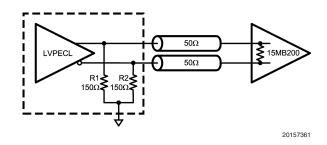


FIGURE 2. DC Coupled LVPECL to LVDS Interface

Figure 2 is a separated π termination scheme for a 3.3 V LVPECL driver. R1 and R2 provides proper DC load for the driver emitter followers, and may be included as part of the

driver device (Note 16). The 15MB200 includes a 100 ohm input termination for the transmission line. The common mode voltage will be at the normal LVPECL levels – around 2 V. This scheme works well with LVDS receivers that have rail-to-rail common mode voltage, V_{CM} , range. Most National Semiconductor LVDS receivers have wide V_{CM} range. The exceptions are noted in devices' respective datasheets. Those LVDS devices that do have a wide V_{CM} range do not vary in performance significantly when receiving a signal with a common mode other than standard LVDS V_{CM} of 1.2 V.

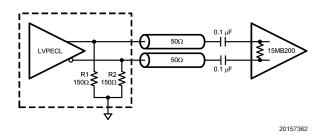


FIGURE 3. AC Coupled LVPECL to LVDS Interface

An AC coupled interface is preferred when transmitter and receiver ground references differ more than 1 V. This is a likely scenario when transmitter and receiver devices are on separate PCBs. *Figure 3* illustrates an AC coupled interface between a LVPECL driver and LVDS receiver. R1 and R2, if not present in the driver device (Note 16), provide DC load for the emitter followers and may range between 140-220 ohms for most LVPECL devices for this particular configuration. The 15MB200 includes an internal 100 ohm resistor to terminate the transmission line for minimal reflections. The signal after ac coupling capacitors will swing around a level set by internal biasing resistors (i.e. fail-safe) which is either V_{DD}/2 or 0 V depending on the actual failsafe implementation. If internal biasing is not implemented, the signal common mode voltage will slowly wander to GND level.

Interfacing LVDS to LVPECL

An LVDS driver consists of a current source (nominal 3.5mA) which drives a CMOS differential pair. It needs a differential resistive load in the range of 70 to 130 ohms to generate LVDS levels. In a system, the load should be selected to match transmission line characteristic differential impedance so that the line is properly terminated. The termination resistor should be placed as close to the receiver inputs as possible. When interfacing an LVDS driver with a non-LVDS receiver, one only needs to bias the LVDS signal so that it is within the common mode range of the receiver. This may be done by using separate biasing voltage which demands another power supply. Some receivers have required biasing voltage available on-chip (V_T, V_{TT} or V_{BB}).

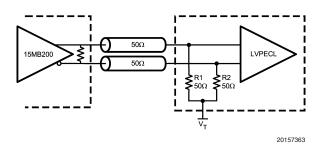


FIGURE 4. DC Coupled LVDS to LVPECL Interface

Figure 4 illustrates interface between an LVDS driver and a LVPECL with a V_T pin available. R1 and R2, if not present in the receiver (Note 16), provide proper resistive load for the driver and termination for the transmission line, and V_T sets desired bias for the receiver.

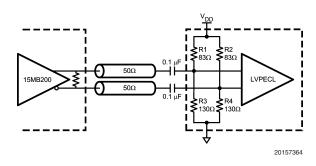


FIGURE 5. AC Coupled LVDS to LVPECL Interface

Figure 5 illustrates AC coupled interface between an LVDS driver and LVPECL receiver without a V_T pin available. The resistors R1, R2, R3, and R4, if not present in the receiver (Note 16), provide a load for the driver, terminate the transmission line, and bias the signal for the receiver.

Note 16: The bias networks shown above for LVPECL drivers and receivers may or may not be present within the driver device. The LVPECL driver and receiver specification must be reviewed closely to ensure compatibility between the driver and receiver terminations and common mode operating ranges.

Packaging Information

The Leadless Leadframe Package (LLP) is a leadframe based chip scale package (CSP) that may enhance chip speed, reduce thermal impedance, and reduce the printed circuit board area required for mounting. The small size and very low profile make this package ideal for high density PCBs used in small-scale electronic applications such as cellular phones, pagers, and handheld PDAs. The LLP package is offered in the no Pullback configuration. In the no Pullback configuration the standard solder pads extend and terminate at the edge of the package. This feature offers a visible solder fillet after board mounting.

The LLP has the following advantages:

- Low thermal resistance
- Reduced electrical parasitics
- Improved board space efficiency
- Reduced package height
- Reduced package mass

For more details about LLP packaging technology, refer to applications note AN-1187, "Leadless Leadframe Package"

